

decoding said composite signal as the bit sequence in the register which contains the largest metric.

2. A method according to claim 1 wherein each metric of said pair that is evaluated for bit $b(x)$ includes an estimated value for bit $b(x)$ times a matched filter output for bit $b(x)$, minus an interference term I_f which contains the matched filter outputs of the $K-1$ bits that immediately follow bit $b(x)$, and minus an interference term I_p which contains $K-1$ bits from the sequence b_s , where K is the number of bit sequences in said composite signal.

3. A method according to claim 2 wherein said term I_f is of the form

$$I_f = \sum_{i=x+1}^{x+k-1} [\text{sign } y(i)] H(x, i),$$

where $y(i)$ is the matched filter output for bit $b(i)$, and $H(x, i)$ is the cross correlation between the spreading codes for bits $b(x)$ and $b(i)$ over the period that bits $b(x)$ and $b(i)$ overlap in said composite signal.

4. A method according to claim 3 wherein said term I_p is of the form

$$I_p = \sum_{i=x-1}^{x-k+1} b(i) H(x, i)$$

where $H(x, i)$ is the cross correlation between the spreading codes for bits $b(x)$ and $b(i)$ over the period that bits $b(x)$ and $b(i)$ overlap in said composite signal.

5. A method according to claim 4 wherein said metric for bit $b(x)$ includes the terms $2b(x)y(x) - b^2(x)W(x) - 2b(x)I_p(x) - 2b(x)I_f(x) - N_o \ln 2$ where $W(x)$ is the power level of bit $b(x)$ and N_o is the power level of any noise in the composite signal.

6. A method according to claim 5 wherein said composite signal is transmitted over said channel via radio waves.

7. A method according to claim 5 wherein said composite signal is transmitted over said channel via an electrical conductor.

8. A decoder for decoding superimposed data bits comprising:

a plurality of filters, which are coupled to receive and filter a composite signal that is formed by coding multiple bit sequences with respective spreading codes and transmitting the coded bit sequences simultaneously and asynchronously over a single

channel, with each filter being matched to a respective one of said spreading codes;

a computing means for evaluating a metric for any bit $b(x)$ in said composite signal which includes an estimated value for bit $b(x)$, an estimate of a sequence of bits b_s which immediately precede bit $b(x)$, and matched filter outputs for several bits in said composite signal which immediately follow bit $b(x)$;

a plurality of register means, each of which is for holding a respective metric and a corresponding bit sequence; and

a control means for repeatedly (a) examining said registers to determine its largest metric M_{max} and corresponding bit sequence b_s , (b) directing said computing means to compute a new pair of metrics for the bit sequences $b_s, b(x)=0$ and $b_s, b(x)=1$, and (c) replacing the register which contains said bit sequence b_s with two registers, one of which contains the bit sequence $b_s, b(x)=0$ and the metric M_{max} plus the above evaluated metric for $b(x)=0$, and the other of which contains the bit sequence $b_s, b(x)=1$ and the metric M_{max} plus the above evaluated metric for $b(x)=1$.

9. A decoder according to claim 8 wherein said computing means evaluates a metric for bit $b(x)$ which includes an estimated value for bit $b(x)$ times a matched filter output for bit $b(x)$, minus an interference term I_f which contains the matched filter outputs of the $K-1$ bits that immediately follow bit $b(x)$, and minus an interference term I_p which contains $K-1$ bits from the sequence b_s , where K is the number of bit sequences in said composite signal.

10. A decoder according to claim 9 wherein said terms I_f and I_p are of the form

$$I_f = \sum_{i=x+1}^{x+k-1} [\text{sign } y(i)] H(x, i)$$

and

$$I_p = \sum_{i=x-1}^{x-k+1} b(i) H(x, i)$$

where $y(i)$ is the matched filter output for bit $b(i)$, and $H(x, i)$ is the cross correlation between the spreading codes for bits $b(x)$ and $b(i)$ over the period that bits $b(x)$ and $b(i)$ overlap in said composite signal.

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